

# DSP

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محاضرة [15]

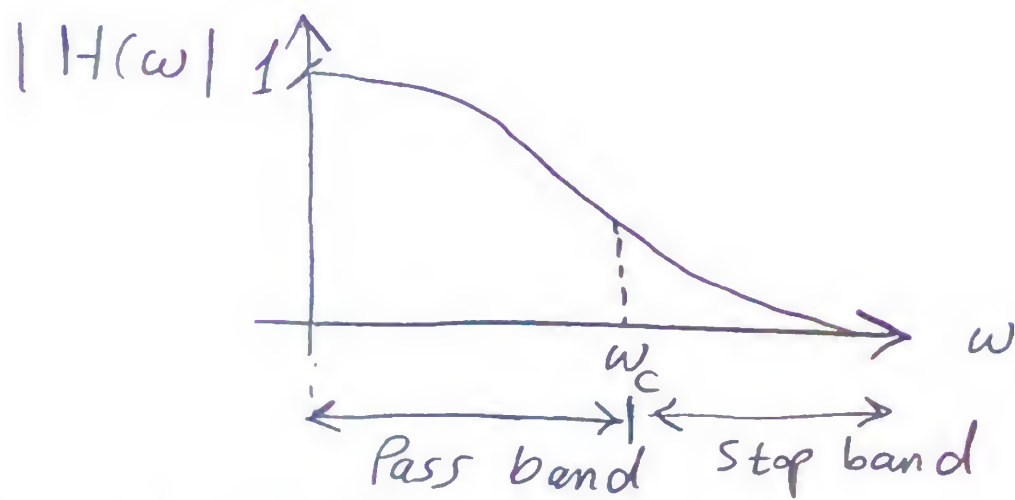
Analog Filter design :-

Digital Filters are discrete time systems that make operations related to Freq. such as :-

- Low pass
- high pass
- Band pass
- Band stop

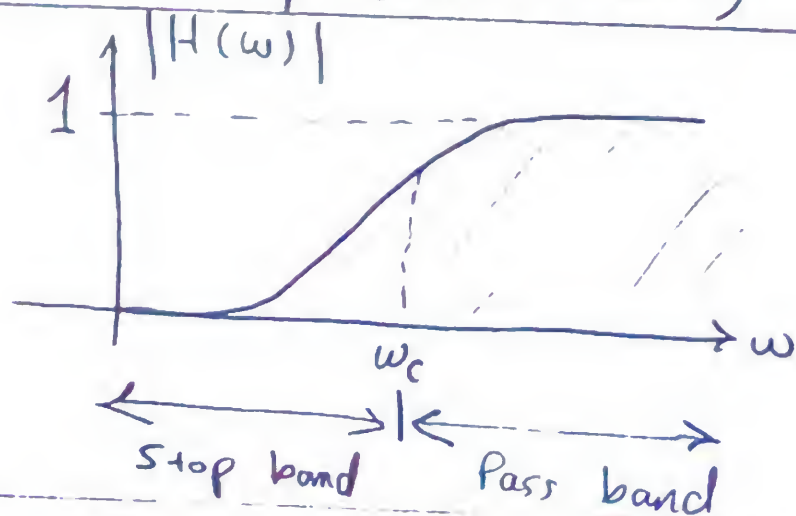
\* types of Filters

① Low pass filter (LPF)

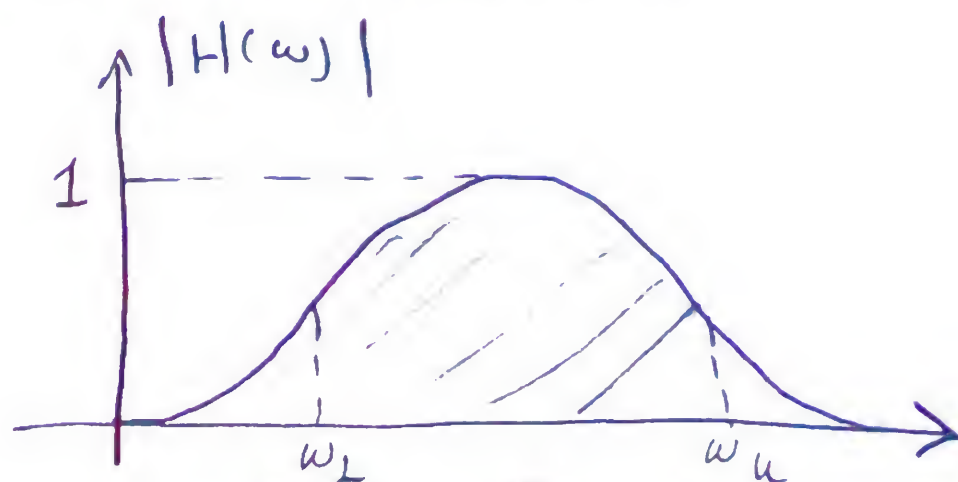


$\omega_c \rightarrow$  cut-off freq (rad/sec)

② High Pass Filter (HPF)



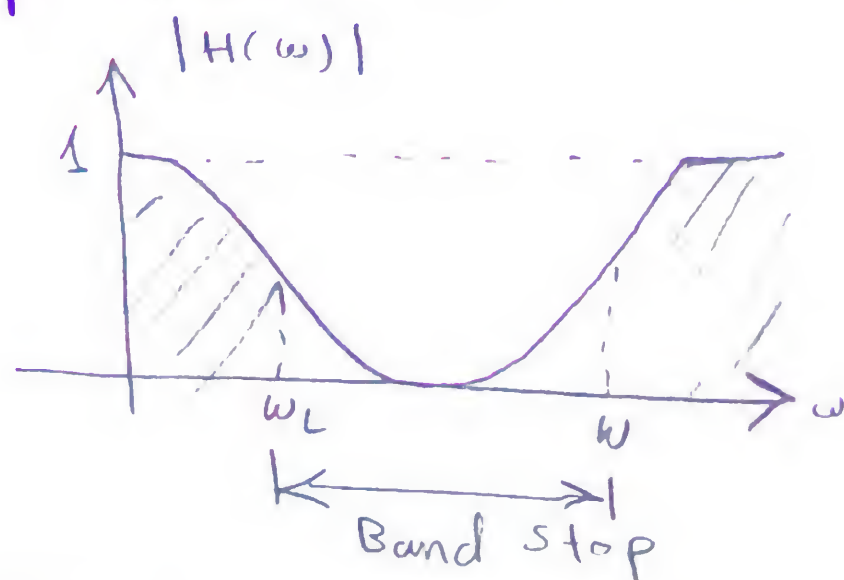
③ Band Pass Filter (BPF)



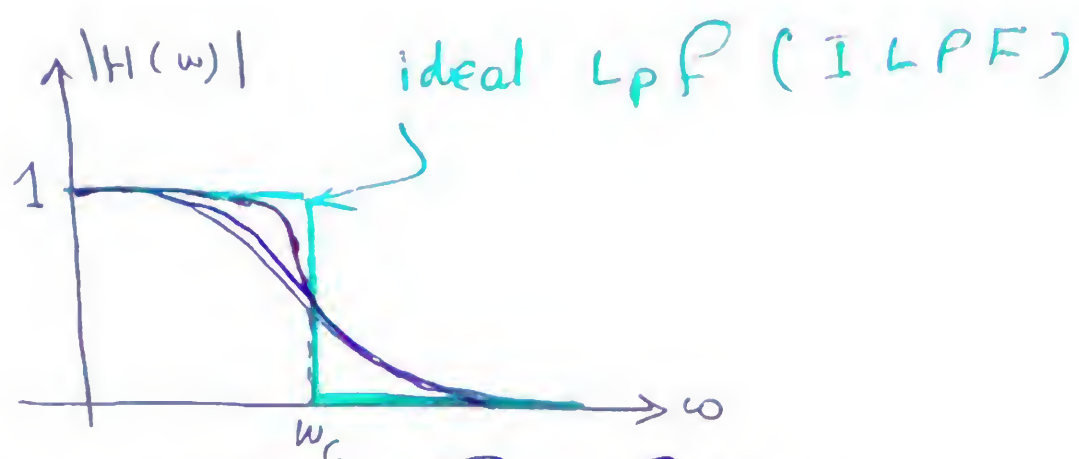
$\omega_L \rightarrow$  lower cut-off frequency.

$\omega_U \rightarrow$  upper cut-off frequency.

#### [4] Band Stop Filter (BSF)

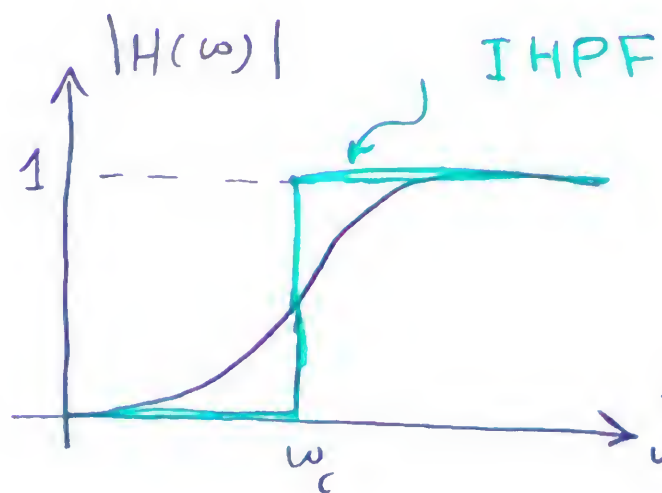


#### [1] LPF



$H(\omega) \Rightarrow$  T. F. for filter

#### [2] HPF



نعمل تقريب لا  
ideal filters

وأعمل Digital Filter  
بنا

# Approximation methods for Analog filter design:-

- ① Butter worth Filter ✓
- ② Elliptic filter
- ③ Bessel Filter
- ④ chebyshev filter



The specs. required for design:-

① cut off freq. ( $\omega_c, \omega_L, \omega_u$ )

② order of the filter

③ type of filter

- LPF
- HPF
- BPF
- BSF

II Butter worth filter Design:-

The design is to find T.F. for the required filter, then we can implement it (H/W or S/W)

\* The Design is made for LPF, then we can convert it to the required type

\* Design for LPF (using Butter worth method):-

- The filter T.F. :-

$$H(s) = \frac{\omega_c^N}{(s - p_0)(s - p_1) \dots (s - p_{N-1})}$$

$N \rightarrow$  the order of the filter

$\omega_c \rightarrow$  cut off frequency (rad/sec)

$$p_k = \omega_c e^{j(N+2k+1)\pi/2N}$$

$$k = 0, 1, \dots, N-1$$

E x1 -

$$\omega_c = 0.725 \text{ rad/sec}, N=2$$

$$H(s) = \frac{(0.725)^2}{(s-p_0)(s-p_1)}$$

$$p_0 = (0.725) e^{j(2+0+1)\frac{\pi}{4}} \angle 135^\circ$$

$$= 0.725 [\cos(135) + j \sin(135)]$$

$$K=0 \rightarrow p_0 = -0.513 + j0.513$$

$$K=1 \rightarrow p_1 = (0.725) e^{j(2+2+1)\frac{\pi}{4}} \angle 225^\circ$$

$$p_1 = -0.513 - j0.513$$

$$H(s) = \frac{(0.725)^2}{\underbrace{(s + 0.513 - j0.513)}_x \underbrace{(s + 0.513 + j0.513)}_y}$$

$$H(s) = \frac{(0.725)^2}{(s + 0.513)^2 + (0.513)^2}$$

$$H(s) = \frac{0.526}{s^2 + 1.026s + 0.526}$$

Normalized LPF = NLPF

A NLPF is an LPF with  $\omega_c = 1 \text{ rad/sec}$

$$N=1 \Rightarrow H(s) = \frac{1}{(s-p_0)}$$

$$p_0 = e^{j(1+0+1)\pi/2} = e^{j\pi} = \cos(\pi) + j \sin(\pi) = -1$$

$$p_k = \omega_c e^{j(N+2k+1)\pi/2N}$$

$$\omega_c = 1$$

$$p_k = e^{j(N+2k+1)\pi/2N}$$



$$H(s)|_{\text{NLPF}} = \frac{1}{s+1}$$

$$N=2$$

$$\Rightarrow H(s)|_{\text{NLPF}} = \frac{1}{(s-p_0)(s-p_1)}$$

$$p_0 = e^{j(2+0+1)\pi/4} \angle 135^\circ$$

$$= \frac{-1}{\sqrt{2}} + j \frac{1}{\sqrt{2}}$$

$$p_1 = e^{j(2+2+1)\pi/4} \angle 225^\circ$$

$$= \frac{-1}{\sqrt{2}} - j \frac{1}{\sqrt{2}}$$

$$H(s)|_{\text{NLPF}} = \frac{1}{\left(s + \frac{-1}{\sqrt{2}} + j \frac{1}{\sqrt{2}}\right) \left(s + \frac{-1}{\sqrt{2}} - j \frac{1}{\sqrt{2}}\right)}$$

$$= \frac{1}{\left(s + \frac{1}{\sqrt{2}}\right)^2 + \frac{1}{2}}$$

$$= \frac{1}{s^2 + \frac{2}{\sqrt{2}}s + \frac{1}{2} + \frac{1}{2}}$$

$$= \frac{1}{s^2 + \sqrt{2}s + 1}$$

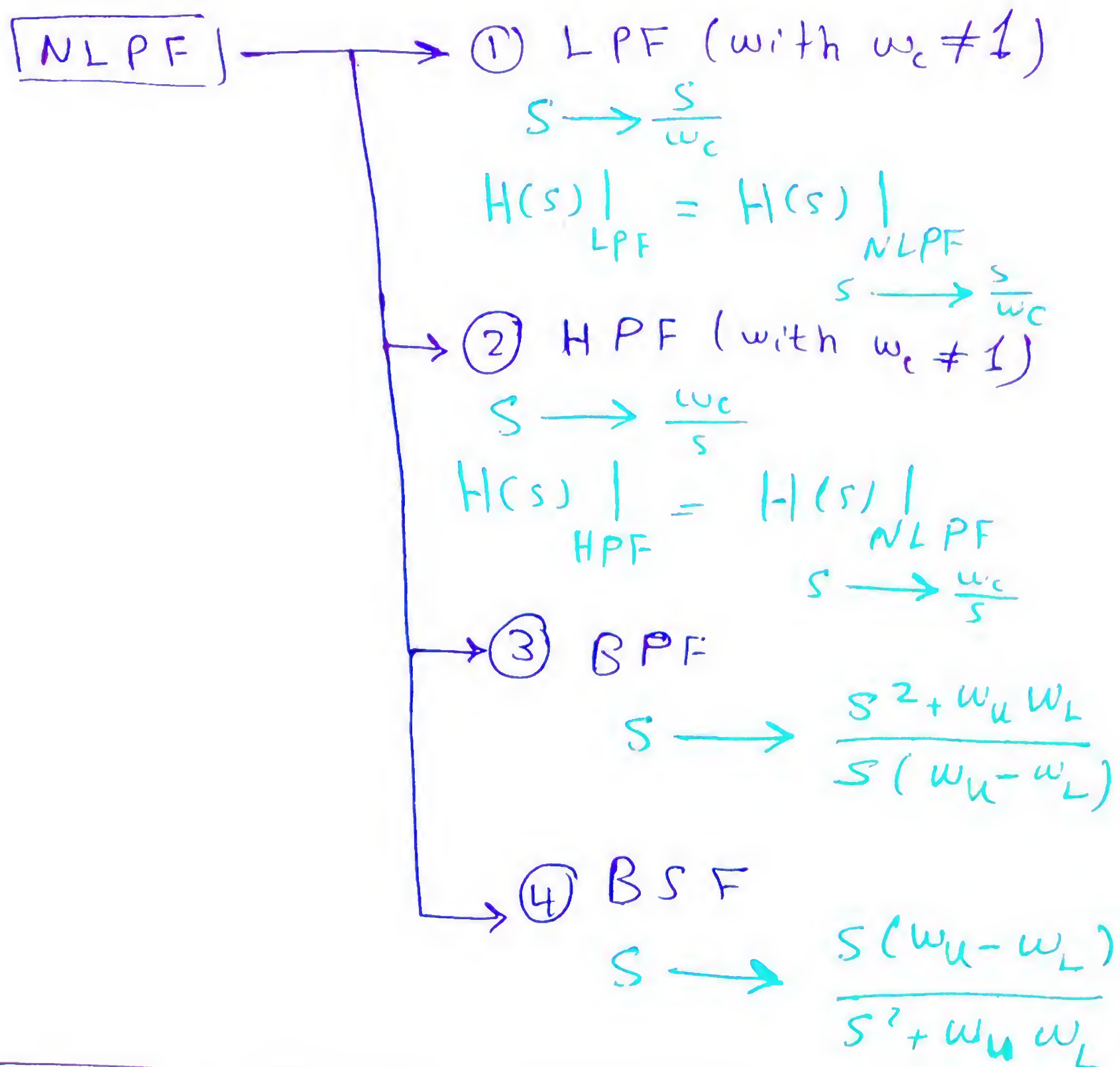
$\therefore$  NLPF (Butter worth)

$$N=1 \longrightarrow H(s) = \frac{1}{s+1}$$

$$N=2 \longrightarrow H(s) = \frac{1}{s^2 + \sqrt{2}s + 1}$$

$$N=3 \longrightarrow H(s) = \frac{1}{(s+1)(s^2+s+1)}$$

Convert from NLPF to any other type :-



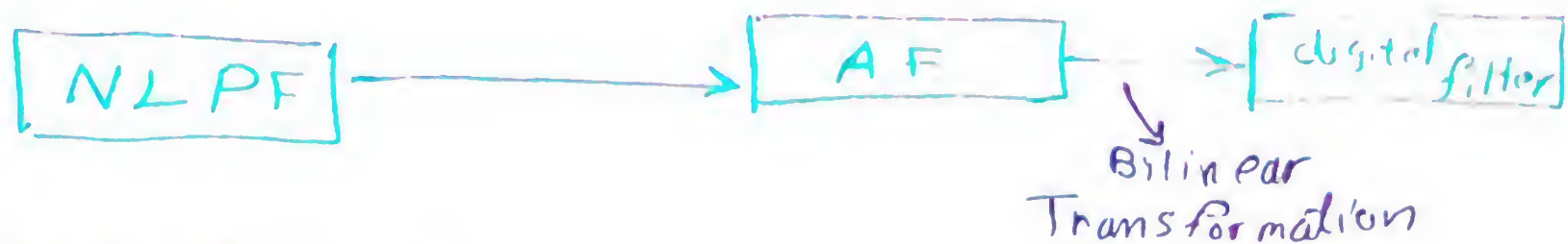
Ex. Design HPF with  $\omega_c = 22 \text{ rad/sec}$  &  $N=3$

$$H(s) \Big|_{\substack{\text{NLPF} \\ N=3}} = \frac{1}{(s+1)(s^2+s+1)}$$

$$\begin{aligned}
 H(s) \Big|_{\substack{\text{HPF} \\ \omega_c = 22 \text{ rad/sec}}} &= \frac{1}{\left(\frac{22}{s} + 1\right) \left(\left(\frac{22}{s}\right)^2 + \frac{22}{s} + 1\right)} \\
 &= \frac{s^3}{(22+s)(22^2 + 22s + s^2)}
 \end{aligned}$$

## Design Digital Filters :—

- \* The design is made in analog domain and then use bilinear transformation to convert the design into Digital domain.



# Bilinear Transformation is a method used to approximate the mapping from  $s$ -domain to  $z$ -domain

$$s = \frac{z}{T} \left( \frac{z-1}{z+1} \right)$$

$$z = e^{Ts} = e^{T/2 \times s}, e^{T/2 \times s}$$

← الإجابات غير مطلوبة

$$= \frac{e^{T/2 s}}{e^{-T/2 s}} = \frac{1 + \frac{T}{2} s}{1 - \frac{T}{2} s}$$

\* From Taylor:

$$e^x = 1 + x + \frac{x^2}{2!} + \frac{x^3}{3!} + \frac{x^4}{4!} + \dots$$

assume  $x \leq 1 \Rightarrow e^x = 1 + x$

$$z = \frac{1 + \frac{T}{2} s}{1 - \frac{T}{2} s}$$

$$z - \frac{T}{2} s z = 1 + \frac{T}{2} s$$

$$z - 1 = \frac{T}{2} s z + \frac{T}{2} s$$

$$z - 1 = (z + 1) \frac{T}{2} s \Rightarrow s = \frac{z}{T} \left( \frac{z-1}{z+1} \right)$$



\* design specs are given in digital domain

$\omega_{cD}, \omega_{uD}, \omega_{LD}, N, \text{Type}, T$   
 $\omega_{cD} \rightarrow$  digital domain

Digital Domain Mapping  $\rightarrow$  Analog Domain

$\omega_c, \omega_u, \omega_L$  Analog domain

$$\omega_A = \frac{2}{T} \tan\left(\frac{\omega_D T}{2}\right)$$

Analog

Ex: Design Bandpass Filter with  $N=1$  with the following specs:

- \*  $f_{LD} = 2.4 \text{ KHz}$  (Lower cutoff freq =  $\checkmark$ )
- \*  $f_{uD} = 2.6 \text{ KHz}$  (upper cutoff freq =  $\checkmark$ )
- \*  $f_s = 8 \text{ KHz}$  (sampling freq =  $\checkmark$ )

$$T = \frac{1}{f_s}$$

radian  $\rightarrow$  rad/sec

□ Convert  $\omega_{LD}, \omega_{uD} \Rightarrow \omega_L, \omega_u$   
 Digital Analog

$$\omega_L = \frac{2}{\frac{1}{8 \times 10^3}} \tan\left(\frac{2.4 \times 10^3 \times 2\pi}{2 \times 8 \times 10^3}\right) \quad T = \frac{1}{8 \times 10^3} \text{ sec}$$

$$= 22022.1107$$

$$\omega_u = \frac{2}{1/8 \times 10^3} \tan\left(\frac{2\pi \times 2.6 \times 10^3}{2 \times 8 \times 10^3}\right)$$

$$= 26109.63 \text{ rad/sec}$$

$$\omega_{LD} = 2\pi f_D$$

$$= 2\pi \times 2.4$$



[2] NLPF with  $N=1$

$$H(s) \Big|_{\substack{\text{NLPF} \\ N=1}} = \frac{1}{s+1}$$

$$H(s) \Big|_{\substack{\text{BPF} \\ \text{with } \omega_L, \omega_U}} = H(s) \Big|_{\substack{\text{NLPF} \\ S \rightarrow \frac{s^2 + \omega_U \omega_L}{s(\omega_U - \omega_L)}}}$$

$$H(s) = \frac{(\omega_U - \omega_L) s}{s^2 + (\omega_U - \omega_L) s + \omega_U \omega_L}$$

$$H(s) \Big|_{\text{ABPF}} = \frac{4087.516 s}{s^2 + 4087.51 s + 574989096.1}$$

[3] Using Bilinear Transformation (B.T)

$$s \rightarrow \frac{z}{T} \left( \frac{z-1}{z+1} \right)$$

$$\Rightarrow H(z) \Big|_{\text{DBPF}} = H(s) \Big|_{\substack{\text{ABPF} \\ S \rightarrow 2f_s \left( \frac{z-1}{z+1} \right)}}$$

$$H(z) = \frac{0.0728z^2 - 0.0728}{z^2 + 0.7118z + 0.854}$$

Ex: design a first order Digital LPF with  $\omega_c = 30\pi \text{ rad/sec}$  and Sampling time  $T = 1/90 \text{ sec}$

given  $\omega_{c0} = 30\pi \text{ rad/sec}$  } Digital domain  
 $T = \frac{1}{90} \text{ sec}$

① obtain  $\omega_c$   
 $\omega_c$  Analog

$$\omega_c = \frac{2}{T} \tan \left( \frac{\omega_c D T}{2} \right)$$

$$= 180 \tan \left( \frac{30\pi}{90 \times 2} \times \frac{180}{\pi} \right) \simeq 103.923 \text{ rad/sec}$$

②  $N=1$ , NLPF

$$H(s) \Big|_{\substack{\text{NLPF} \\ N=1}} = \frac{1}{s+1}$$

$$H(s) \Big|_{\substack{\text{LPF} \\ N=1 \\ \omega_c = 103.923}} = H(s) \Big|_{\substack{\text{NLPF} \\ s \rightarrow \frac{s}{\omega_c}}} = \frac{1}{\frac{s}{\omega_c} + 1} = \frac{\omega_c}{s + \omega_c}$$

$$= \frac{103.923}{s + 103.923}$$

③ Using B.T.

$$H(z) = H(s) \Big|_{\text{LPF}} = \frac{103.923}{180 \left( \frac{z-1}{z+1} \right) + 103.923}$$

$$s \rightarrow \frac{z}{T} \left( \frac{z-1}{z+1} \right)$$

$$H(z) \Big|_{\text{DLPF}} = \frac{0.366(z+1)}{z - 0.2679}$$